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1. Current through voltage divider is

$$I = \frac{V_{cc} - V_{EE}}{R_1 + R_2}$$

$$V_{out} = V_{cc} - IR_1 = \frac{V_{cc}(R_1 + R_2) - V_{cc}R_1 + V_{EE}R_1}{R_1 + R_2}$$

$$= \frac{V_{cc}R_2 + V_{EE}R_1}{R_1 + R_2}$$

$$V_{out}(R_1 + R_2) = V_{cc}R_2 + V_{EE}R_1$$

$$(V_{cc} - V_{out})R_2 = (V_{out} - V_{EE})R_1$$

The output impedance is the parallel combination of R_1 and R_2 :

$$Z = \frac{R_1 R_2}{R_1 + R_2}$$

$$\frac{1}{Z} = \frac{1}{R_1} + \frac{1}{R_2}$$

$$\frac{1}{R_1} = \left(\frac{V_{out} - V_{EE}}{V_{cc} - V_{out}} \right) \frac{1}{R_2} = \frac{1}{Z} - \frac{1}{R_2}$$

$$\text{So } \frac{1}{Z} = \left(\frac{V_{out} - V_{EE} + V_{cc} - V_{out}}{V_{cc} - V_{out}} \right) \frac{1}{R_2}$$

$$Z = \left(\frac{V_{cc} - V_{out}}{V_{cc} - V_{EE}} \right) R_2$$

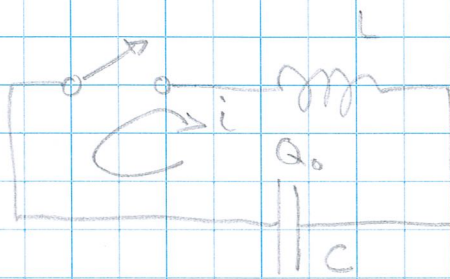
$$R_2 = Z \left(\frac{V_{cc} - V_{EE}}{V_{cc} - V_{out}} \right)$$

$$R_1 = \left(\frac{V_{cc} - V_{out}}{V_{out} - V_{EE}} \right) R_2 = Z \left(\frac{V_{cc} - V_{EE}}{V_{out} - V_{EE}} \right)$$

$$\begin{aligned}\text{Thus, } R_1 &= (1\text{ k}\Omega) \left(\frac{5\text{ V} - (-5\text{ V})}{0.5\text{ V} - (-5\text{ V})} \right) \\ &= (1\text{ k}\Omega) \left(\frac{10\text{ V}}{5.5\text{ V}} \right) \\ &= 1.82\text{ k}\Omega\end{aligned}$$

$$\begin{aligned}R_2 &= (1\text{ k}\Omega) \left(\frac{5\text{ V} - (-5\text{ V})}{5\text{ V} - 0.5\text{ V}} \right) \\ &= (1\text{ k}\Omega) \left(\frac{10\text{ V}}{4.5\text{ V}} \right) \\ &= 2.22\text{ k}\Omega\end{aligned}$$

2.



$$-L \frac{di}{dt} - \frac{1}{C} \left(Q_0 + \int_0^t i(t) dt \right) = 0$$

$$\frac{d^2 i}{dt^2} + \frac{1}{LC} i(t) = 0$$

$$\frac{d^2 i}{dt^2} + \omega^2 i(t) = 0$$

$$\omega = \sqrt{\frac{1}{LC}}$$

$$i(t) = A \sin \omega t + B \cos \omega t$$

$$\text{At } t=0, i(t)=0 \text{ so } B=0.$$

$$\text{Also, at } t=0, \frac{di}{dt} = A\omega \cos \omega t$$

$$-A\omega - \frac{Q_0}{LC} = 0$$

$$\text{So } A = -\frac{Q_0}{LC\omega} = -\frac{Q_0}{\sqrt{LC}}$$

$$i(t) = -\frac{Q_0}{\sqrt{LC}} \sin \omega t$$

(a) Its a low pass filter.

3.

$$V_{in} - I_1 R - (I_1 - I_2) Z_c = 0$$

$$-I_2 R - I_2 Z_c - (I_2 - I_1) Z_c = 0$$

$$\begin{pmatrix} R+Z_c & -Z_c \\ -Z_c & R+2Z_c \end{pmatrix} \begin{pmatrix} I_1 \\ I_2 \end{pmatrix} = \begin{pmatrix} V_{in} \\ 0 \end{pmatrix}$$

$$I_2 = \frac{\begin{vmatrix} R+Z_c & V_{in} \\ -Z_c & 0 \end{vmatrix}}{\begin{vmatrix} R+Z_c & -Z_c \\ -Z_c & R+2Z_c \end{vmatrix}}$$

$$= \frac{V_{in} Z_c}{(R+Z_c)(R+2Z_c) - Z_c^2}$$

$$(R+Z_c)(R+2Z_c) - Z_c^2$$

$$V_{out} = I_2 Z_c = \frac{V_{in} Z_c^2}{R^2 + 3RZ_c + Z_c^2}$$

$$Z_c = \frac{-i}{\omega C}$$

$$V_{out} = V_{in} \left(\frac{1}{-1 + \omega^2 R^2 C^2 + 3i\omega RC} \right)$$

$$\left| \frac{V_{out}}{V_{in}} \right| = \frac{1}{((1 + \omega^2 R^2 C^2)^2 + 9\omega^2 R^2 C^2)^{1/2}}$$

-3db at

$$\left| \frac{V_{out}}{V_{in}} \right| = \frac{1}{\sqrt{2}}$$

$$\omega^4 R^4 C^4 + 11\omega^2 R^2 C^2 + 1 = 2$$

$$\omega^2 = \frac{-11R^2 C^2 \pm \sqrt{(11R^2 C^2)^2 + R^4 C^4}}{2R^4 C^4} = \frac{1}{R^2 C^2} \left(\frac{\sqrt{122} - 11}{2} \right)$$

(6)

As $\omega \rightarrow \infty$,

$$A = \left| \frac{V_{out}}{V_{in}} \right| \rightarrow \frac{1}{\omega^2 R^2 C^2}$$

$$G \rightarrow 20 \log_{10} A = -40 + \text{const.}$$

Gain falls off at 40 dB per decade.

4.

$$R + \frac{Z_0 + R}{3} = Z_0$$

$$3R + Z_0 + R = 3Z_0$$

$$4R = 2Z_0$$

$$R = \frac{1}{2} Z_0$$

In general, $R + \frac{Z_0 + R}{n} = Z_0$

$$nR + Z_0 + R = nZ_0$$

$$(n+1)R = (n-1)Z_0$$

$$R = Z_0 \left(\frac{n-1}{n+1} \right)$$

when $n = 2$, $R = \frac{Z_0}{3}$

when $n = 3$, $R = \frac{Z_0}{2}$

$n = 4$, $R = \frac{3Z_0}{5}$