

## Sec. 9.5 Multipole expansion for a source in a waveguide

For starters, I give an overview of results from Sec 8.12B. Specifically, we can expand  $\vec{E}, \vec{H}$  inside, in terms of separately propagating waves along  $\pm \hat{z}$ , namely

$$\vec{E} = \vec{E}^{(+)} + \vec{E}^{(-)}$$

$$\vec{H} = \vec{H}^{(+)} + \vec{H}^{(-)}$$

where the individual pieces have a mode expansion:

$$\vec{E}^{(\pm)} = \sum_{\lambda} A_{\lambda}^{(\pm)} \vec{E}_{\lambda}^{(\pm)}$$

$$\vec{H}^{(\pm)} = \sum_{\lambda} A_{\lambda}^{(\pm)} \vec{H}_{\lambda}^{(\pm)}$$

For any given problem, the expansion coefficients  $A_{\lambda}^{(\pm)}$  can often be found using the mode orthonormality relations, e.g. 8.131-134. Note that the TRANSVERSE FIELDS obey an orthonormality relation. I simply state these relations here, without proof.

The full fields have the following form for each mode:

$$\vec{E}_\lambda^{(\pm)}(x, y, z) = [\vec{E}_\lambda(x, y) + \vec{E}_{z\lambda}(x, y)] e^{\pm ik_\lambda z}$$

and

$$\vec{H}_\lambda^{(\pm)}(x, y, z) = [\pm \vec{H}_\lambda(x, y) + \vec{H}_{z\lambda}(x, y)] e^{\pm ik_\lambda z}$$

and the transverse component orthonormality relations are simply, normalizing the field modes to be REAL:

$$\int \vec{E}_\lambda \cdot \vec{E}_\mu da = \delta_{\lambda\mu}$$

and

$$\int \vec{H}_\lambda \cdot \vec{H}_\mu da = \frac{1}{Z_\lambda^2} \delta_{\lambda\mu}$$

and the time-averaged power flow obeys

$$\frac{1}{2} \int (\vec{E}_\lambda \times \vec{H}_\mu) \cdot \hat{z} da = \frac{1}{2 Z_\lambda} \delta_{\lambda\mu}$$

The longitudinal components also obey orthonormality relations,

$$\int E_{z\lambda} E_{z\mu} da = -\frac{\gamma_\lambda^2}{k_\lambda^2} \delta_{\lambda\mu} \quad \text{TM modes}$$

$$\int H_{z\lambda} H_{z\mu} da = -\frac{\gamma_\lambda^2}{k_\lambda^2 Z_\lambda^2} \delta_{\lambda\mu} \quad \text{TE modes}$$

The explicit normalized modes for a rectangular wave guide are in Eqs 8.135-8.136

Another aside from Chap. 8, pp 392-4 derives two key formulas needed in Chap. 9:

(1) A localized current source  $\vec{J}(\vec{x})$  in the waveguide produces fields whose expansion coefficients are

$$A_{\lambda}^{(\pm)} = -\frac{Z_{\lambda}}{2} \int_V \vec{J} \cdot \vec{E}_{\lambda}^{(\mp)} d^3x \quad \leftarrow \text{note } \mp \text{ order!} \quad 8.146$$

AND

(2) A localized aperture in a waveguide generates additional fields in addition to an unperturbed propagating field  $\vec{E}$ , whose expansion coefficients are:

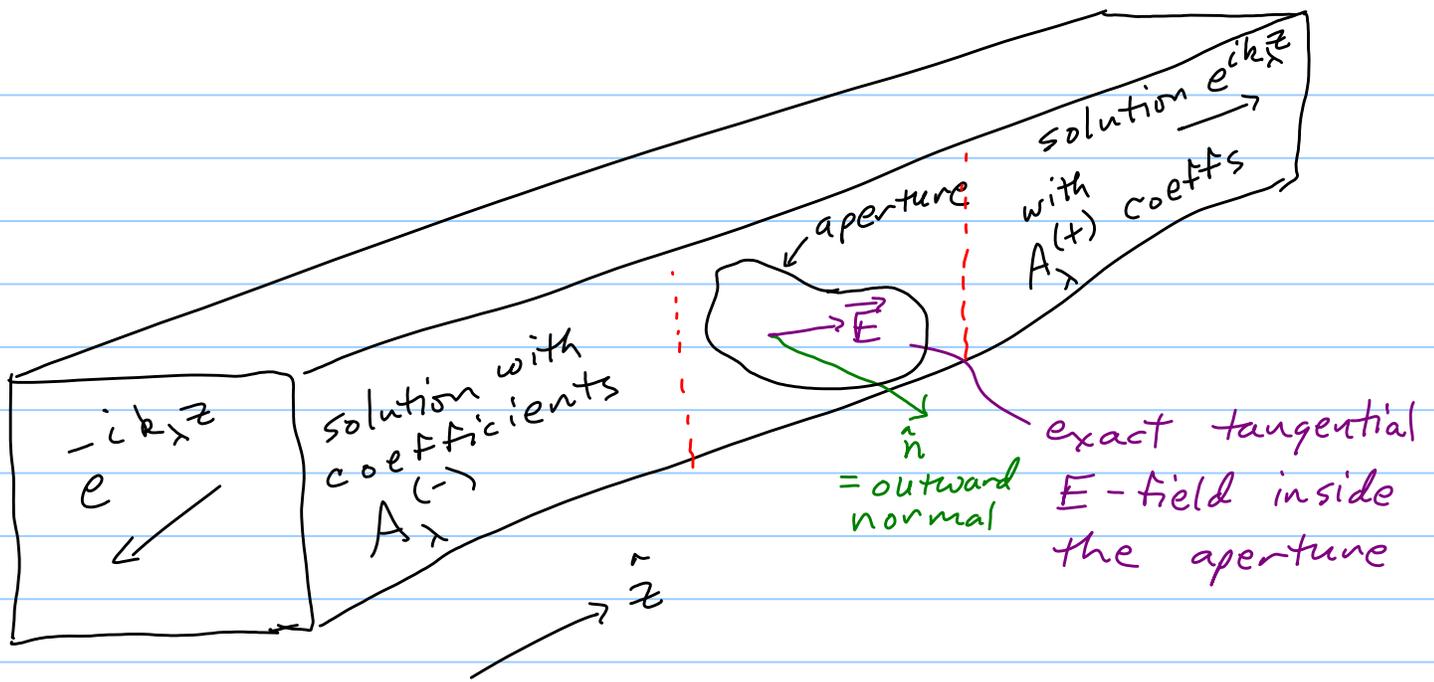
$$A_{\lambda}^{(\pm)} = \frac{Z_{\lambda}}{2} \int_{\text{aperture}} (\vec{E} \times \vec{H}_{\lambda}^{(\mp)}) \cdot \hat{n} da \quad 8.147$$

Where  $\vec{E} =$  exact tangential  $\vec{E}$ -field in the aperture,

and

$\hat{n} =$  OUTWARDLY-directed normal unit vector

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Now return to Sec 9.5, the multipole exp.

for a small source inside a waveguide

↑ means small compared to the distance over which the fields vary

This "smallness" suggests that we can expand the  $\vec{E}_{\lambda}^{(\pm)}$  into a Taylor series.

- use repeated index summation convention

- omit the labels  $\lambda$  and  $(\pm)$  for simplicity:

$$\Rightarrow \int \vec{J} \cdot \vec{E} d^3x = \int \underbrace{J_{\alpha}(\vec{x})}_{\text{term 1}} \left[ \underbrace{E_{\alpha}(0) + (\vec{x} \cdot \nabla) E_{\alpha}(0) + \dots}_{\text{term 2}} \right] d^3x$$

$$\text{term 1} = \vec{E}(0) \cdot \int \vec{J} d^3x = -i\omega \vec{P} \cdot \vec{E}(0) \quad \text{where}$$

$$\vec{P} = \frac{i}{\omega} \int \vec{J}(\vec{x}) d^3x = \text{source electric dipole moment}$$

Eq. 9.14

term 2 =  $\int d^3x \left\{ \mathcal{J}_\alpha \chi_\beta \frac{\partial E_\alpha(0)}{\partial \chi_\beta} \right\} \leftarrow$  now analyze this as we did Eq. 9.30

$$= \int d^3x \left\{ \frac{1}{4} (\mathcal{J}_\alpha \chi_\beta - \mathcal{J}_\beta \chi_\alpha) \left( \frac{\partial E_\alpha(0)}{\partial \chi_\beta} - \frac{\partial E_\beta(0)}{\partial \chi_\alpha} \right) + \frac{1}{2} (\mathcal{J}_\alpha \chi_\beta + \mathcal{J}_\beta \chi_\alpha) \frac{\partial E_\alpha(0)}{\partial \chi_\beta} \right\}$$

gives a magnetic dipole term

gives an electric quadrupole contribution

or term 2 =  $i\omega \vec{m} \cdot \vec{B}(0) - \frac{i\omega}{6} \left[ \nabla \cdot (\vec{Q} \cdot \vec{E}) \right]_{\vec{r} \rightarrow 0}$

So finally, this gives

$$A_\lambda^{(\pm)} = \frac{i\omega Z_\lambda}{2} \left\{ \vec{P} \cdot \vec{E}_\lambda^{(\mp)}(0) - \vec{m} \cdot \vec{B}_\lambda^{(\mp)}(0) + \frac{1}{6} \left[ \nabla \cdot (\vec{Q} \cdot \vec{E}_\lambda^{(\mp)}) \right]_{\vec{r} \rightarrow 0} + \dots \right\} \quad (9.69)$$

Similarly, if there is a localized aperture

in a waveguide carrying  $\Sigma M$ -fields, we can expand in terms of the effective dipole moments associated with the aperture, (9.71)

$$A_\lambda^{(\pm)} = \frac{i\omega Z_\lambda}{4} \left[ \vec{P}_{\text{eff}} \cdot \vec{E}_\lambda^{(\mp)}(0) - \vec{m}_{\text{eff}} \cdot \vec{B}_\lambda^{(\mp)}(0) + \dots \right]$$

where the "effective dipole moments" are

$$\vec{P}_{\text{eff}} = \epsilon \hat{n} \int \vec{x} \cdot \vec{E}_{\text{tangential}} da$$

and

$$\vec{M}_{\text{eff}} = \frac{2}{i\mu\omega} \int \hat{n} \times \vec{E}_{\text{tangential}} da$$

and  $\vec{E}_{\text{tangential}}$  = exact tangential  $\vec{E}$ -field  
in the aperture